



n.v. delta elektronika

nieuwe boogerdstraat 2 zierikzee holland telefoon (01110) 2734

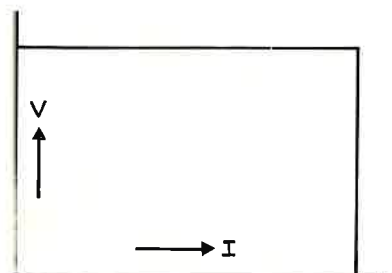


REGULATED POWER SUPPLY D 030-3

up to serial number 8172

0 – 30 V, 0 – 3 A

- Voltage and current regulated.
- Silicon transistors.
- 10-turn potentiometers for voltage- and current controls.
- Voltage and current programming.
- Output terminals at the front and at the back.
- Parallel and series connection permitted at all load conditions.
- Bench model can also be used for 19-inch rack mounting.
- Constant voltage regulation changes sharply into constant current regulation.
- Can be short circuited.



Remote programming	Voltage and current can be programmed with variable resistors of 0-5000 Ohm which can be connected at the back side.
Remote sensing	The output voltage can be regulated at a load point remote from the power supply by means of two extra wires.
Voltage regulation	The output voltage changes less than + or - 3 mV for a AC line voltage variation of + or - 10 % and less than + or - 10 mV for a maximum load change.
Current regulation	The output current changes less than + or - 3 mA for a AC line voltage variation of + or - 10 % and less than + or - 3 mA for a maximum load change.
Ripple voltage	At voltage regulation the ripple voltage is less than 0,1 mV r.m.s.
Ripple current	At current regulation the ripple current is less than 0,5 mA r.m.s.
Output impedance	At voltage regulation the output impedance is less than 3 milli-ohms for slow load variations and less than 200 milli-ohms for sine shape load variations up to 100 kHz. At current regulation the output impedance is larger than 10 kilo-ohms for slow load variations.
Temperature coefficient	The temperature dependence of the constant voltage is less than + or - 3 mV per °C. The temperature dependence of the constant current is less than + or - 3 mA per °C.
Meters	Voltage meter 0 - 30 V accuracy 1.5 %. Current meter 0 - 3 A accuracy 1.5 %.
Cooling	Heat removal is by natural convection. One should take care that the air can flow freely vertically through the cooling fins.
Grounding	The output terminals are isolated from the case.
Models	Suffix A is a bench model. Suffix B is uncased. Two power supplies D 030-3 B can be screwed together and be used in a 19-inch rack with two mounting pieces H2.
Input voltage	220 V AC 50 Hz
Dimensions and weight	D 030-3 A 221 x 150 x 250 mm 11,4 KG. D 030-3 B 215 x 133 x 250 mm 10,0 KG. (width x height x depth).

Das stabilisierte Netzgerät D 030-3 kann als Spannungsquelle mit konstanter Spannung und begrenztem Strom, oder als Stromquelle mit konstantem Strom und begrenzter offener Spannung benutzt werden.

Fernstabilisierung

Die Ausgangsspannung kann, mit Hilfe von 2 Extra Leitungen, an einem Verbraucherpunkt vom Netzgerät entfernt, stabilisiert werden (Abb.1).

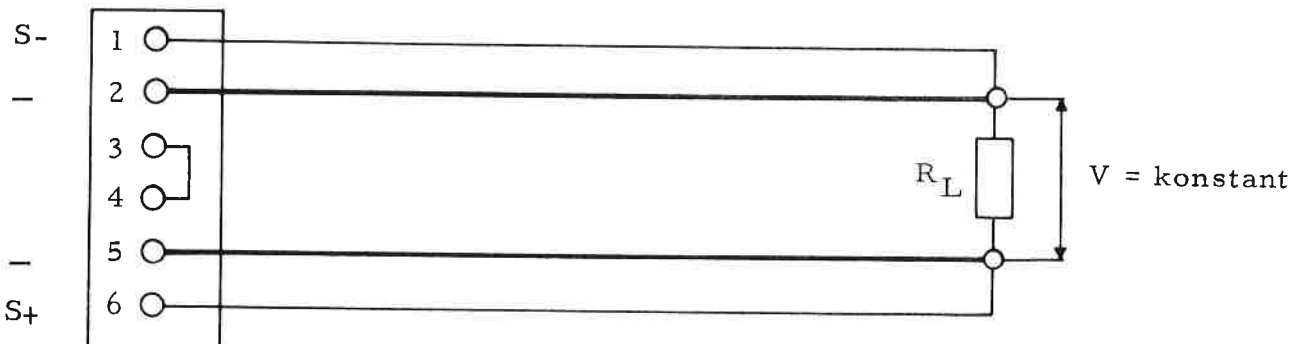


Abb. 1

Die Kurzschluss-Verbindungen zwischen den Klemmen S+ und + und den Klemmen S- und -, an der Rückseite, sind in diesem Falle zu entfernen.

Ein Spannungsabfall bis zu 1,5 Volt kann in jeder Leitung kompensiert werden. Abhängig von der Ausgangsspannung und dem Strom kann man in vielen Fällen grössere Spannungsabfälle kompensieren.

Bei Fernstabilisierung wird die Induktivität der Verbindungsleitungen zum Verbraucher nicht kompensiert.

Um einen niedrigen Scheinwiderstand der Spannungsquelle am Verbraucherpunkt zu erhalten, wäre ein Kondensator parallel zu den Belastungsklemmen nützlich.

Um die Induktivität zu verringern, sind die Drähte zu verdrehen. Die Fühler-Drähte können ebenso verdreht werden.

Umgebungstemperatur

Eine Umgebungstemperatur bis 40°C ist gestattet wenn ununterbrochen belastet mit 3 A und bis 70°C wenn ununterbrochen belastet mit 1,5 A.

Wird das Gerät in Gestellaufbau verwendet, so ist für ausreichende Luftkühlung und Abfuhr der erzeugten Wärme zu sorgen.

Schaltungsbeschreibung

Zur Beschreibung der Schaltung ist ein einfaches Schaltbild beigelegt (Abb. 2).

Die Stabilisierung besteht aus zwei Teilen.

Eine schnelle Stabilisierung mit Silizium Transistoren und eine langsame Vor-Stabilisierung mit Thyristoren (Silicon controlled Rectifiers).

Vor-Stabilisierung:

Die Vor-Stabilisierung mit Thyristoren wird benutzt um die Verlustleistung im Transistor T niedrig zu halten.

Zu diesem Zweck wird die Spannung über T unabhängig von der Eingangs- und Ausgangsspannung konstant gehalten.

Die Spannung über T wird mit einem Teil der Spannungsreferenz verglichen (Die Spannung über R_1 wird zur Vereinfachung nicht betrachtet).

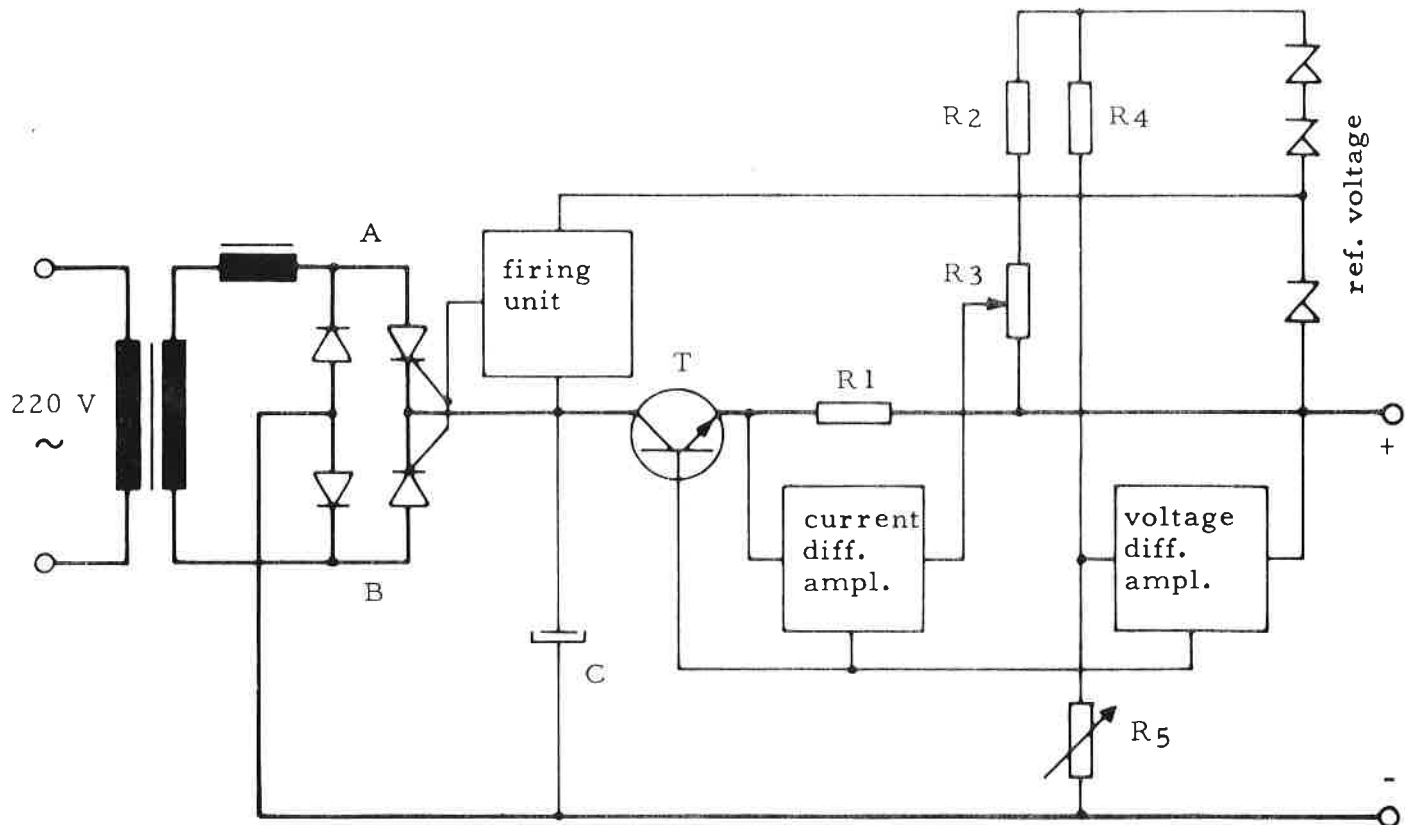


Fig. 2 Simplified Circuitdiagram D 050-10
D 030- 3

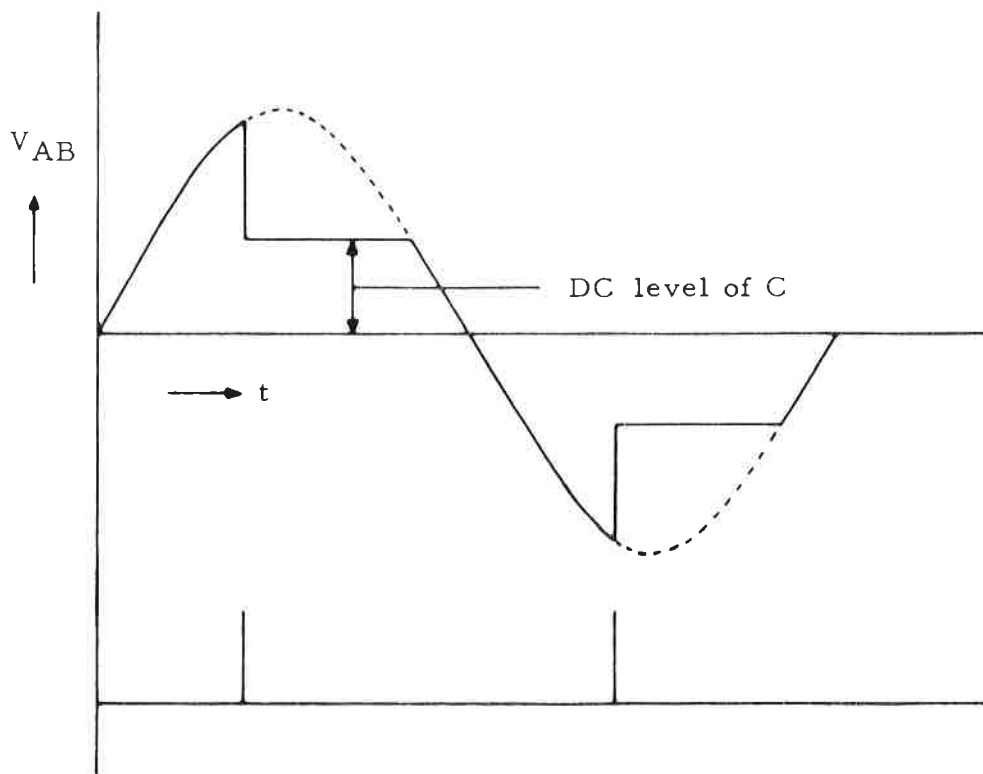


Fig. 3

The error voltage is converted into a time shift of the firing pulses of the controlled rectifiers, with regard to the zero crossings of the sine wave input voltage.

This causes a voltage change across the capacitor C in order to keep the voltage across T constant.

At a load current of 3 Amp. the voltage across T is about 3 Volts.

So the dissipation of T is than 9 Watts.

Without pre-regulation the dissipation would be about 90 Watts at low output voltages.

Constant voltage regulation:

A comparison bridge is formed with the resistors R4, R5, the reference voltage and the output voltage.

After amplification the error voltage of the bridge drives transistor T. At the condition of balance the output voltage is practically proportional to R5.

As long as the constant voltage regulation is active, the constant current regulation is inoperative, because one of the transistors of the current error amplifier is blocked.

Constant current regulation:

At constant current regulation the voltage drop across R1, which is proportional to the output current, is compared with the part of the reference voltage across R3.

The error voltage is amplified and drives transistor T in such a way that the voltage across R1 is kept constant and this means that the output current is constant.

The constant current is adjustable with potentiometer R3.

As long as the constant current regulation is active, the constant voltage regulation is inoperative, because one of the transistors of the voltage error amplifier is blocked.

The position of the crossover point of constant voltage regulation and constant current regulation depends on the settings of the voltage and current controls.

In fig. 4 the crossover point is drawn.

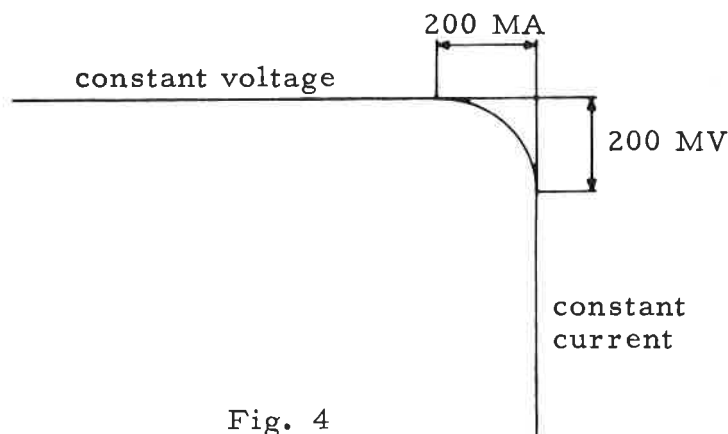


Fig. 4

R (Ohm)

1 =	1,2	k	5W 5%
2 =	10	k	var.
3 =	1	k	var.
4 =	560		$\frac{1}{2}$ W 2% MF
5 =	15	k	$\frac{1}{2}$ W 2% MF
6 =	4,7	k	$\frac{1}{2}$ W 2% MF
7 =	ARW		
8 =	10	k	$\frac{1}{2}$ W 2% MF
9 =	1,2	k	$\frac{1}{2}$ W 2% MF
10 =	1,5	k	$\frac{1}{2}$ W 5%
11 =	1	k	$\frac{1}{2}$ W 5%
12 =	33	k	$\frac{1}{2}$ W 5%
13 =	68	k	1W 2% MF
14 =	10	k	$\frac{1}{2}$ W 2% MF
15 =	15	k	$\frac{1}{2}$ W 2% MF
16 =	1	k	$\frac{1}{2}$ W 5%
17 =	18	k	$\frac{1}{2}$ W 5%
18 =	4,7	k	$\frac{1}{2}$ W 2% MF
19 =	ARW		
20 =	100	k	$\frac{1}{2}$ W 5%
21 =	8,2	k	$\frac{1}{2}$ W 5%
22 =	1	k	$\frac{1}{2}$ W 5%
23 =	1	k	$\frac{1}{2}$ W 5%
24 =	100		$\frac{1}{2}$ W 5%
25 =	4,7	k	$\frac{1}{2}$ W 5%
26 =	470		$\frac{1}{2}$ W 5%
27 =	100		$\frac{1}{2}$ W 5%
28 =	330		$\frac{1}{2}$ W 5%
29 =	1,8	k	$\frac{1}{2}$ W 5%
30 =	1	k	var.
31 =	330		$\frac{1}{2}$ W 5%
32 =	330		$\frac{1}{2}$ W 5%
33 =	3,3	k	$\frac{1}{2}$ W 5%
34 =	8,2	k	$\frac{1}{2}$ W 5%
35 =	10	k	var.
36 =	47	k	$\frac{1}{2}$ W 5%
37 =	15	k	$\frac{1}{2}$ W 5%
38 =	470	k	$\frac{1}{2}$ W 5%
39 =	1	m	var.
40 =	3,9	k	$\frac{1}{2}$ W 5%
41 =	ARW		
42 =	30	k	$\frac{1}{2}$ W 2% MF
43 =	1	k	var.
44 =	1,2	k	9W 5%
45 =	10		$\frac{1}{2}$ W 5%
46 =	0,6		dr. gew.
47 =	5	k	var. 10 sl. potm.
48 =	5	k	var. 10 sl. potm.
49 =	1,2	k	9W 5%
50 =	330		5W 5%
51 =	150		5W 5%

C (microfarad)

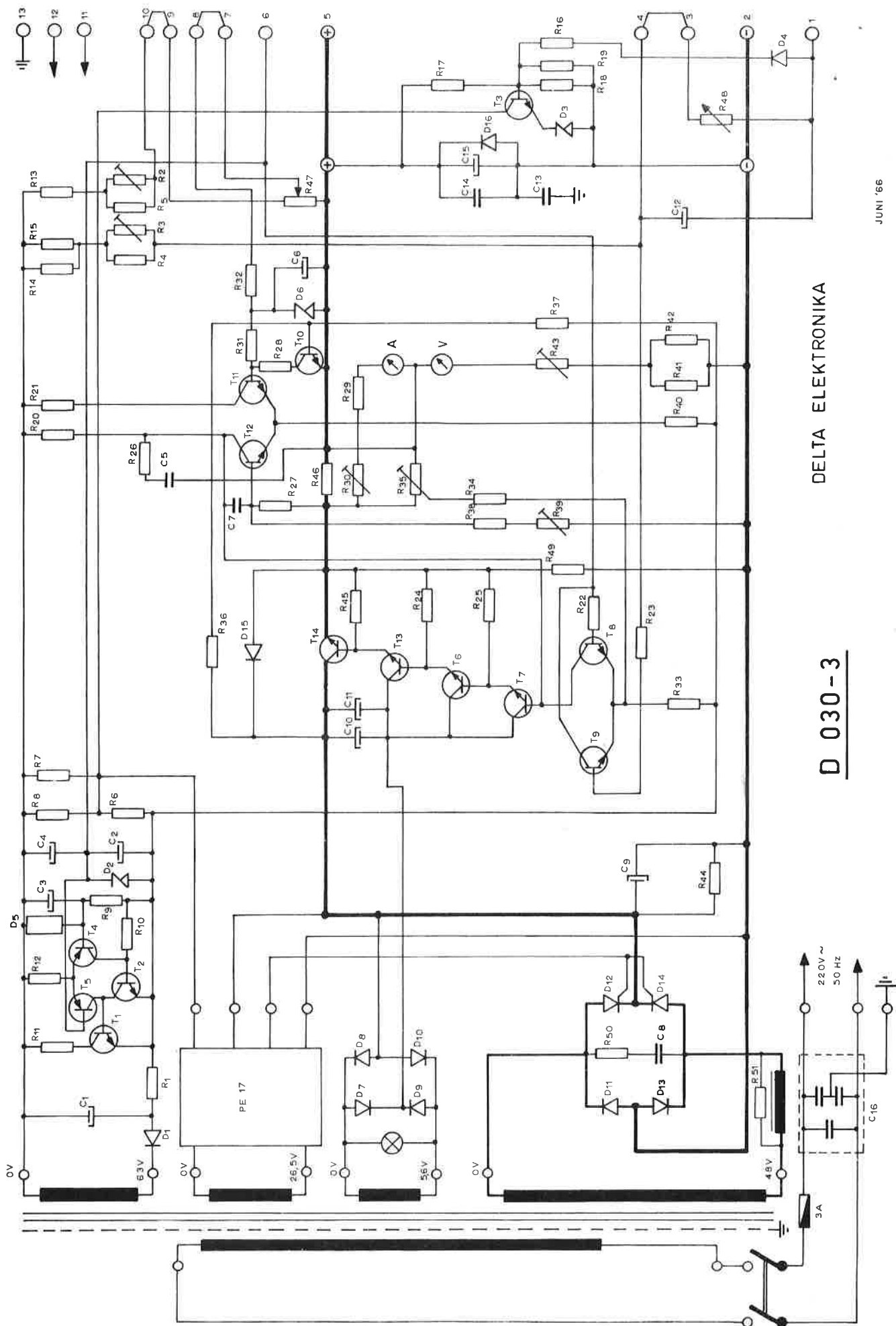
1 =	100	100 V	T 1 =	40348	RCA
2 =	25	15 V	2 =	2N3704	TI
3 =	10	100 V	3 =	40348	RCA
4 =	25	70 V	4 =	2N3703	TI
5 =	0,00047	400 V	5 =	2N3703	TI
6 =	10	100 V	6 =	40348	RCA
7 =	0,00047	400 V	7 =	40348	RCA
8 =	1	250 V	8 =	2N3704	TI
9 =	10.000	70 V	9 =	2N3704	TI
10 =	250	15 V	10 =	2N3704	TI
11 =	250	15 V	11 =	2N3704	TI
12 =	10	100 V	12 =	2N3704	TI
13 =	0,1	630 V	13 =	2N3055	RCA
14 =	0,47	250 V	14 =	2N3055	RCA
15 =	500	70 V			
16 =	0,2 + 2 x 0,005	250 V			

D 1 =	TS -2	DI
2 =	ZG 6,8	Intermetall
3 =	ZG 6,8	Intermetall
4 =	OA 202	Philips
5 =	RE 38	Delta
6 =	ZG 6,8	Intermetall
7 =	TS 05	DI
8 =	TS 05	DI
9 =	TS 05	DI
10 =	TS 05	DI
11 =	40209	RCA
12 =	2N3668	RCA
13 =	40209	RCA
14 =	2N3668	RCA
15 =	MR 1031 B	Motorola
16 =	40209	RCA

MF = Metaalfilmweerstand
Metalfilmresistor
Metalfilmwiderstand

ARW = Afregelweerstand
Calibration resistor
Abregelwiderstand

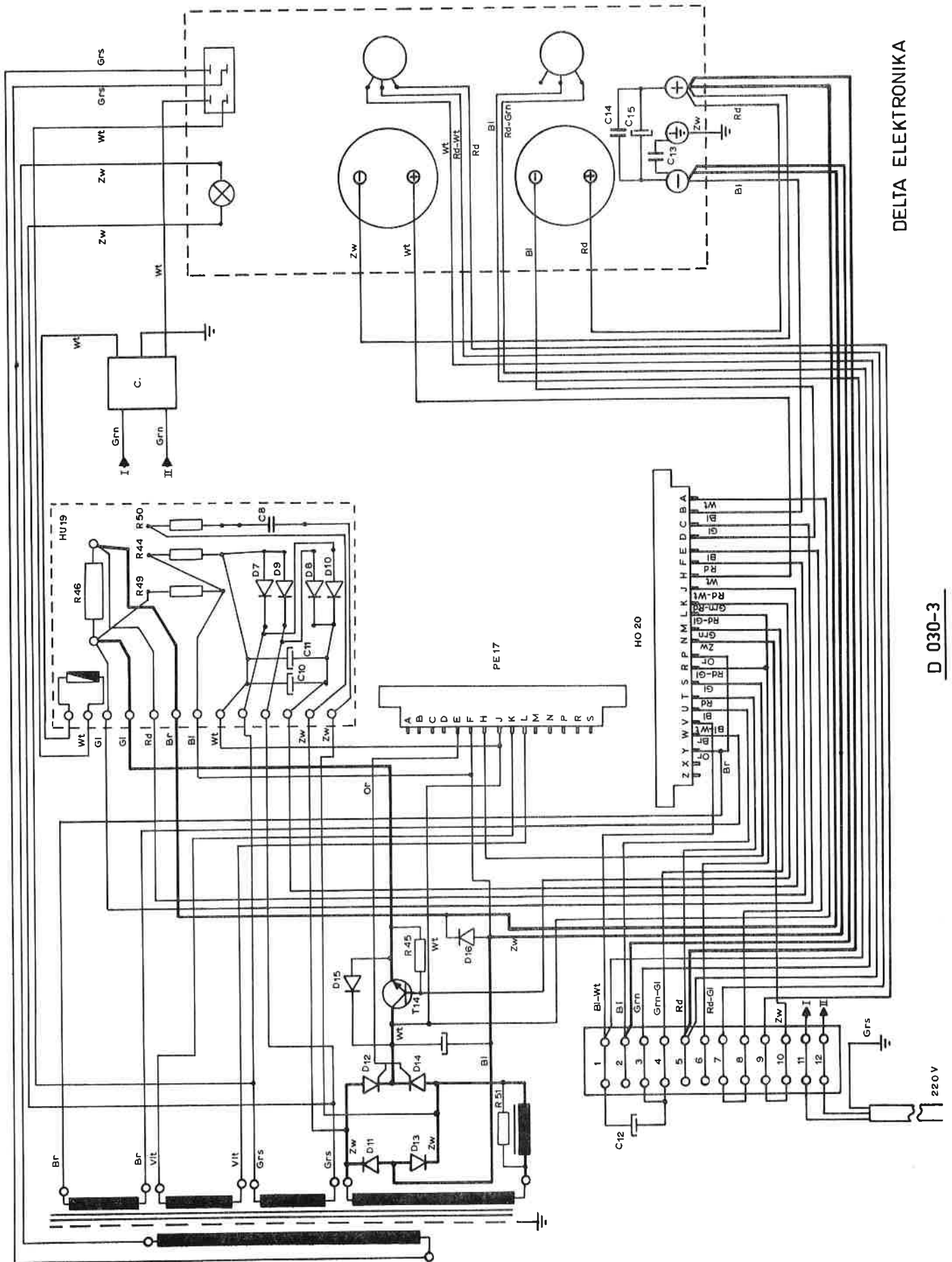
L = Liliput telefonlampe
6 V 0,04 A
Fabr. Taunuslicht



DELTA ELEKTRONIKA

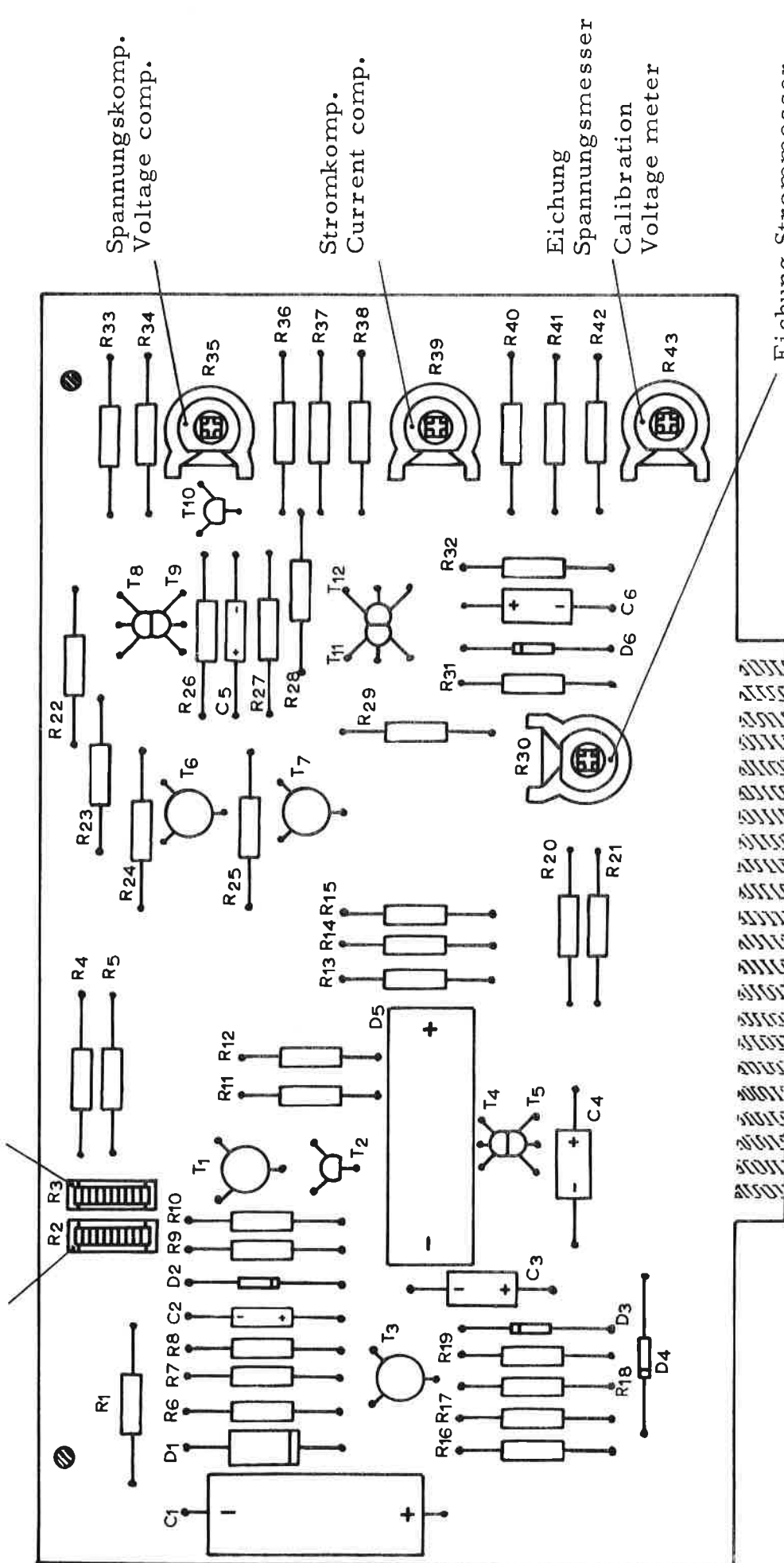
D 030-3

JUNI '66

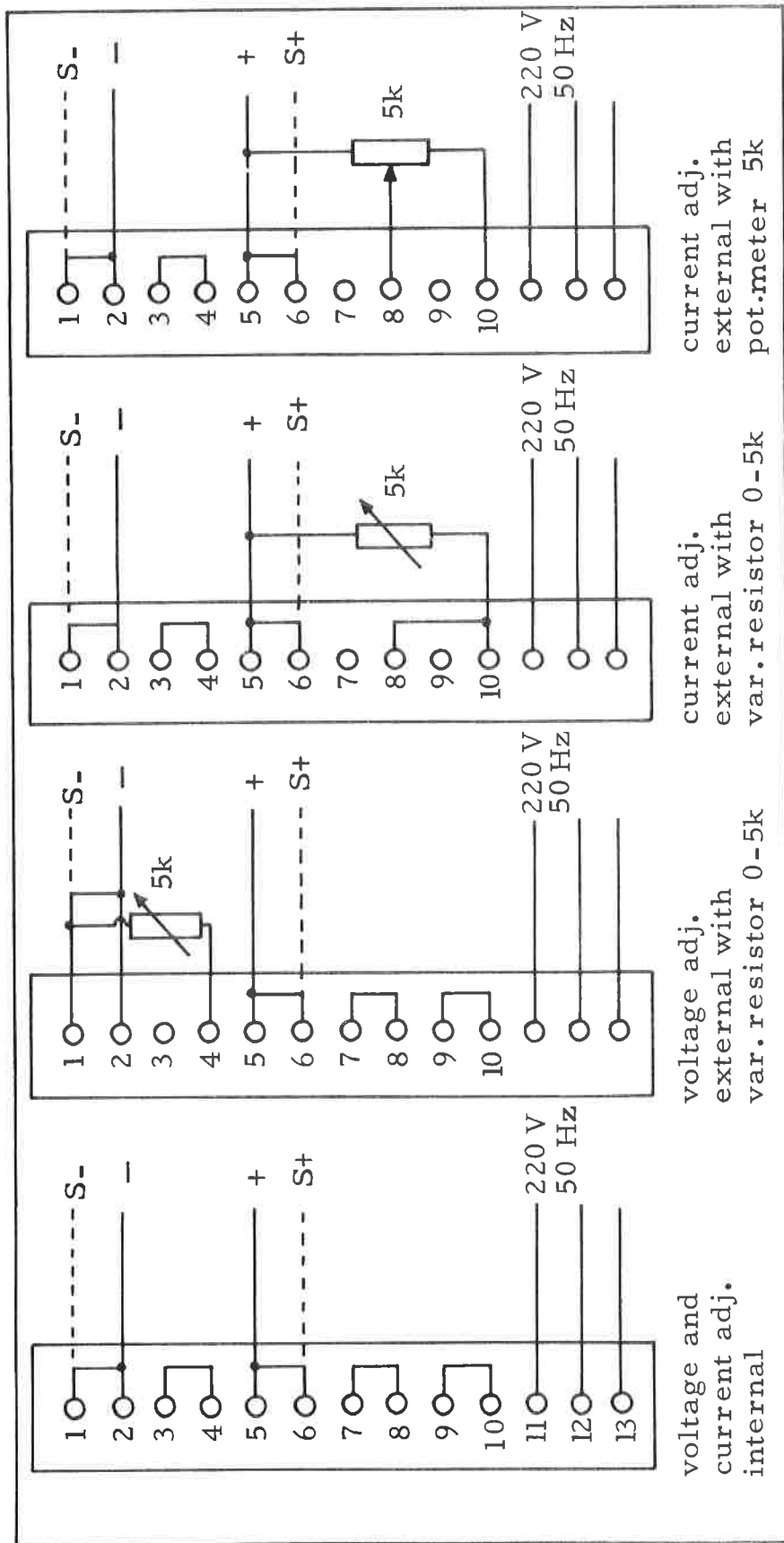


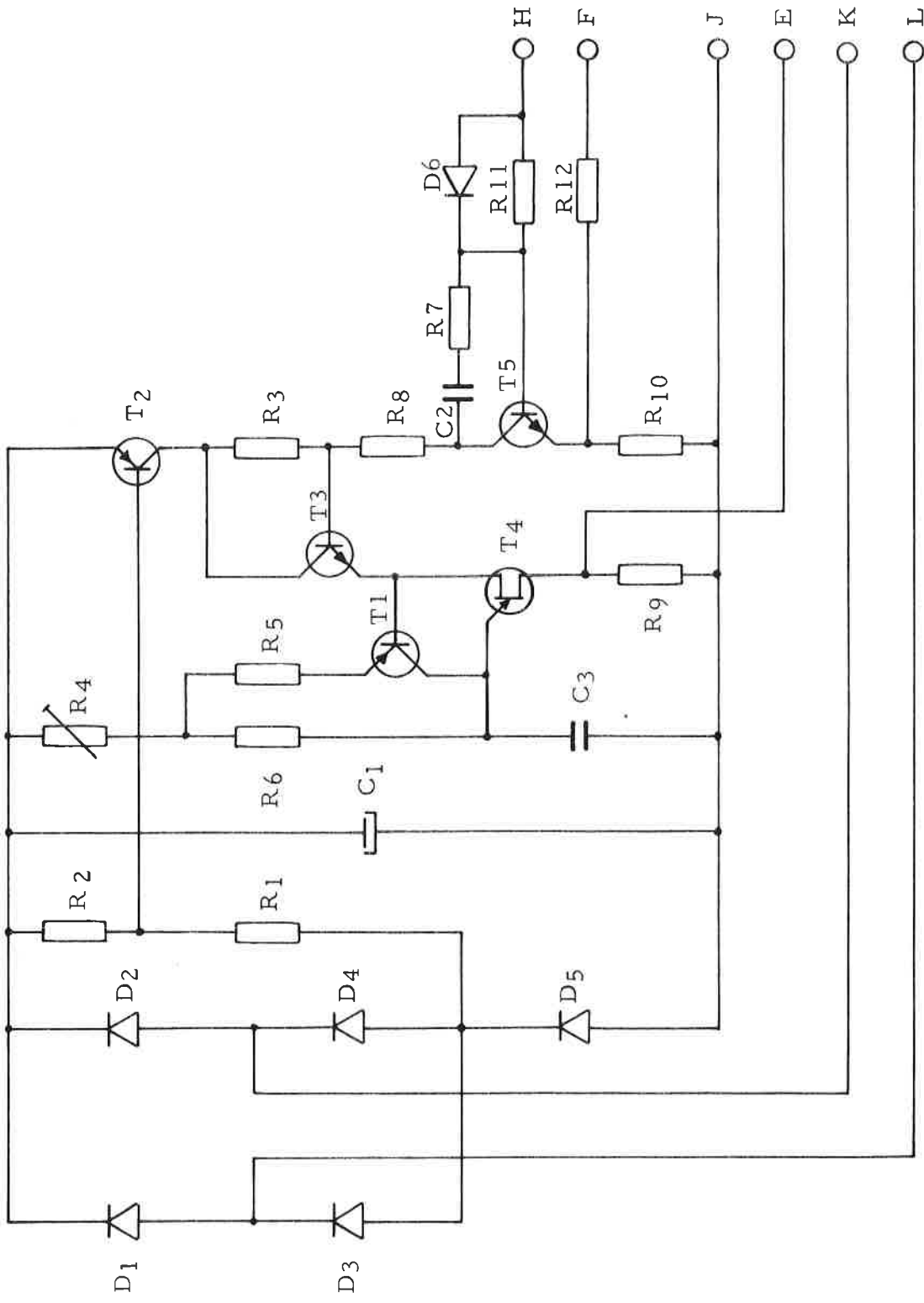
Strombereich
Current range

Spannungsbereich
Voltage range



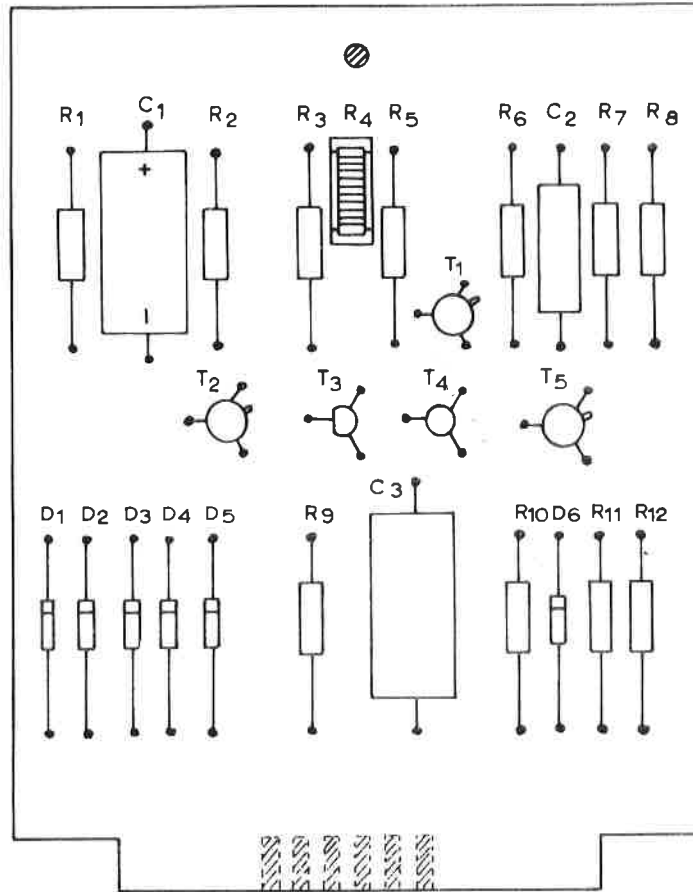
HO 20





D 030-3

PE 17



PE 17

R (Ohm)

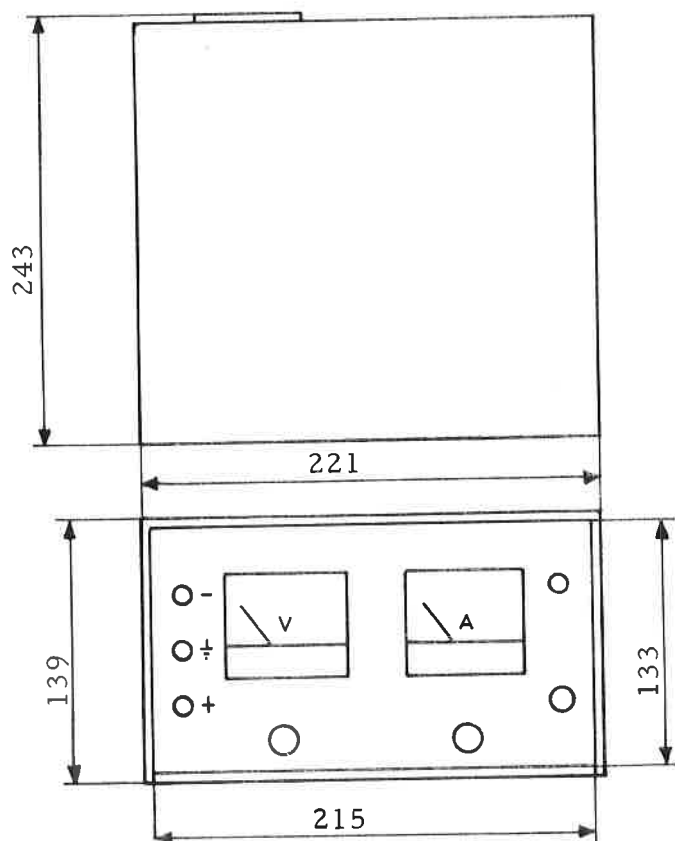
1 = 4,7 k	$\frac{1}{2}$ W	5%
2 = 4,7 k	$\frac{1}{2}$ W	5%
3 = 4,7 k	$\frac{1}{2}$ W	2% MF
4 = 10 k	var.	
5 = 33 k	$\frac{1}{2}$ W	2%
6 = 56 k	$\frac{1}{2}$ W	2%
7 = 22 k	$\frac{1}{2}$ W	5%
8 = 2,7 k	$\frac{1}{2}$ W	2% MF
9 = 27 k	$\frac{1}{2}$ W	5%
10 = 2,7 k	$\frac{1}{2}$ W	2% MF
11 = 10 k	$\frac{1}{2}$ W	5%
12 = 22 k	$\frac{1}{2}$ W	5%

C (microfarad)

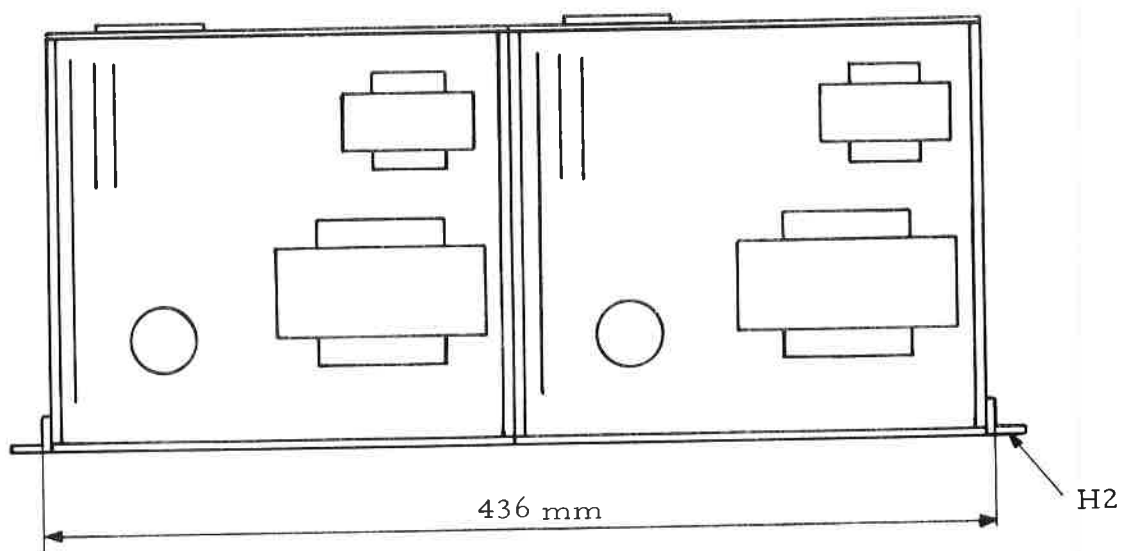
1 = 50	70 V
2 = 0,1	250 V
3 = 0,22	63 V

T 1 =	2N4037	RCA
2 =	2N4037	RCA
3 =	2N3704	TI
4 =	2N2646	GE
5 =	40348	RCA

D 1 =	TS 05	DI
2 =	TS 05	DI
3 =	TS 05	DI
4 =	TS 05	DI
5 =	TS 05	DI
6 =	OA 202	Philips



D 030-3 A



2 x D 030-3 B



REGULATED POWER SUPPLY D 030-3

0 – 30 V, 0 – 3 A

from serial number 8173

DESCRIPTION

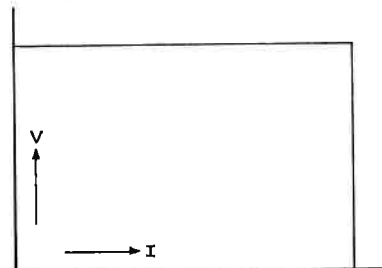
The power supply D 030-3 can be used as a constant voltage source with a limited current or as a constant current source with a limited open voltage. The change of mode occurs sharply at the crossing of the voltage and current settings.

A preregulator with silicon controlled rectifiers keeps the rectified voltage in accordance with the output voltage.

This means low dissipation in the transistors of the series regulator, so that no blower is needed for cooling.

The preregulator causes no interference on the mains.

The power supply is protected against any overload condition.



CONSTANT VOLTAGE OPERATION

- | | |
|---------------------------|---|
| Voltage control | 10-turn potentiometer, range 0-30 V. |
| Remote programming | The voltage can be programmed by an external variable resistor of 0-5000 Ohm. Input on the rear panel. |
| Remote sensing | Separate amplifier terminals enable the output voltage to be regulated at a remote load point, using two sensing leads. |
| Voltage regulation | 3 mV for a + or - 10 % AC input voltage variation.
10 mV for a 100 % load change. |
| Temp. coeff. | Less than 3 mV per °C. |
| Ripple voltage | 0.1 mV r.m.s. |

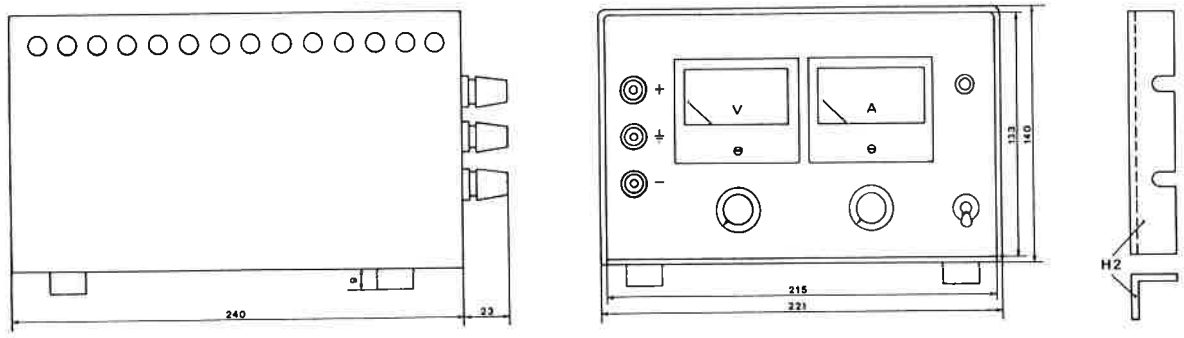
Output impedance	Maximum 0.1 Ohm up to 100 kHz load frequency.
Recovery time	20 micro seconds for recovery to within 30 mV of steady state voltage after a step load change from 10 % to 100 %.

CONSTANT CURRENT OPERATION

Current control	10-turn potentiometer, range 0-3 A.
Remote programming	The current can be programmed by an external variable resistor of 0-5000 Ohm. Input on the rear panel.
Current regulation	3 mA for a + or - 10 % AC input voltage variation. 3 mA for a maximum output voltage swing.
Temp. coeff.	Less than 3 mA per °C.
Ripple current	0.5 mA r.m.s.

REMAINING SPECIFICATIONS

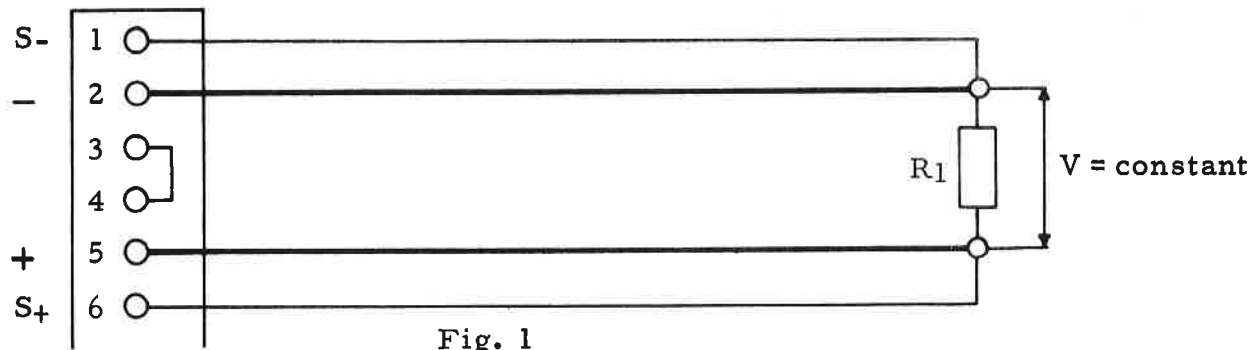
Input voltage	110-120-220-240 V, 50 Hz.
Parallel and series connection	Special design enables parallel and series operation without any precaution.
Ambient temp.	- 20 to + 45 °C.
Output terminals	On front and rear panel, isolated from the case. Maximum voltage between output terminals and case 500 V.
Rack mounting	Two uncased units can be mounted side by side and with the addition of two H2 brackets can be inserted in a 19" rack. For ordering uncased units add B to type number (D 030-3 B).
Cooling	By natural convection cooling. The air must flow freely through the vertical heat sink for effective cooling.
Meters	Voltage meter 0-30 V, accuracy 1.5 %. Current meter 0- 3 A, accuracy 1.5 %.
Finish	Light gray front panel with dark blue case.
Weight and size	11.4 kg, 221 x 140 x 240 mm.



The power supply D 030-3 can be used as a source of constant voltage with a limited current, or as a source of constant current with a limited open voltage.

Remote sensing

The output voltage may be regulated at a load point remote from the power supply by means of two extra wires (fig. 1).



The shorting links between the terminals S+ and + and between the terminals S- and - at the back side of the power supply have to be disconnected in this case.

A voltage drop up to 1.5 Volts in each connecting wire can be compensated. Depending on the output voltage and current it is often possible to compensate for still larger voltage drops.

Remote sensing does not compensate for the inductance of the load connecting wires.

To establish a low source impedance at the load a capacitor bypass directly at the load terminals is useful.

To minimize the inductance the load wires should be twisted together. The sensing wires can also be twisted together.

Ambient temperature

The maximum allowed ambient temperature is 40°C when the load current is 3 A continuously and 70°C at 1,5 A.

When mounted in a rack or cabinet one should care for sufficient ventilation to remove the dissipated heat.

Circuit description

To explain the circuit a simplified circuit diagram is drawn (fig. 2).

The regulation consist of two parts: A fast regulation with silicon transistors and a slower pre-regulation with silicon controlled rectifiers.

Pre-regulation:

The switching pre-regulation with controlled rectifiers is used to keep the dissipation in the pass transistor T low.

For this purpose the voltage across T is kept constant, independent of the input and output voltage.

The voltage across T is compared with a part of the reference voltage (the voltage across R_1 is neglected for simplification).

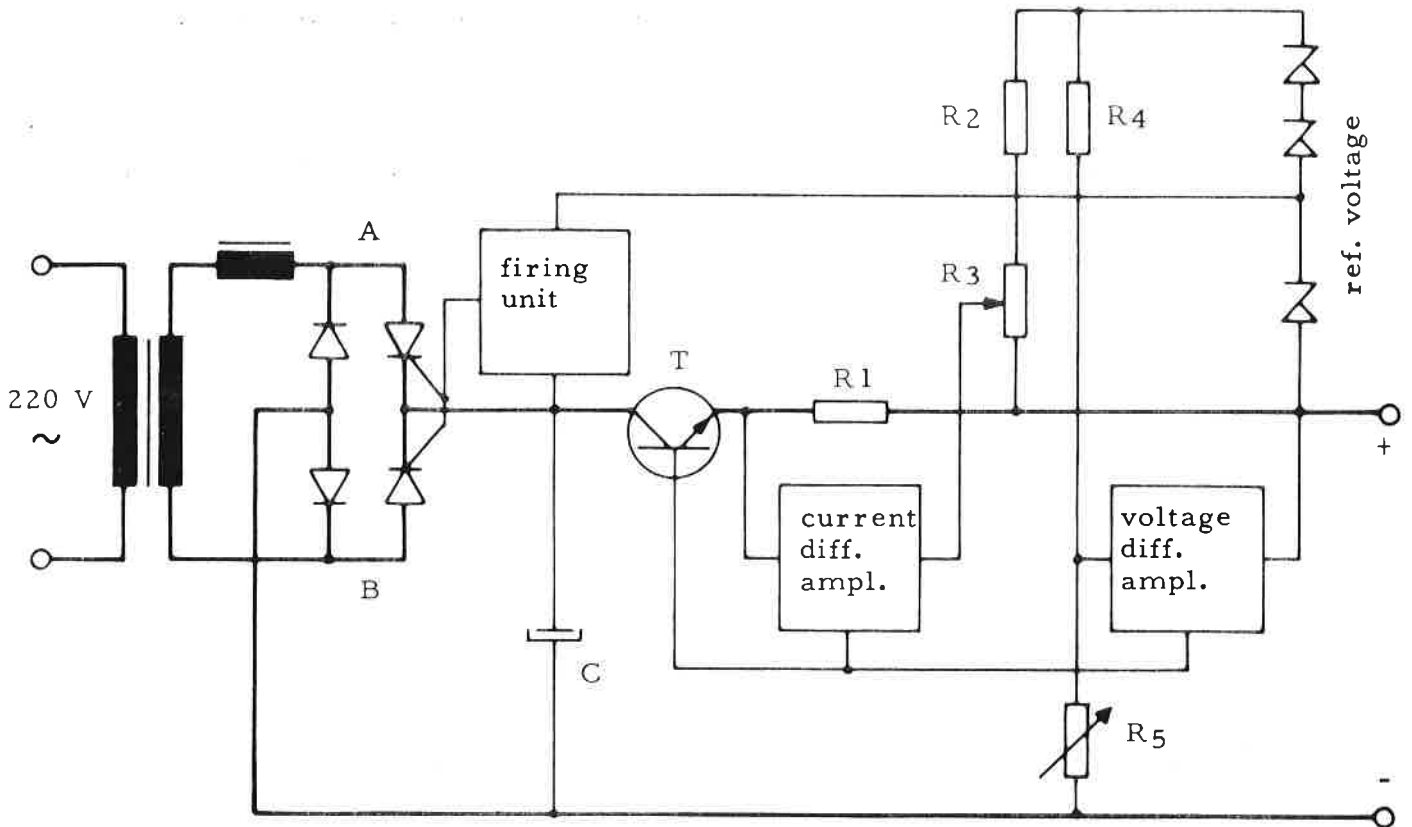


Fig. 2 Simplified Circuitdiagram D 050-10
D 030- 3

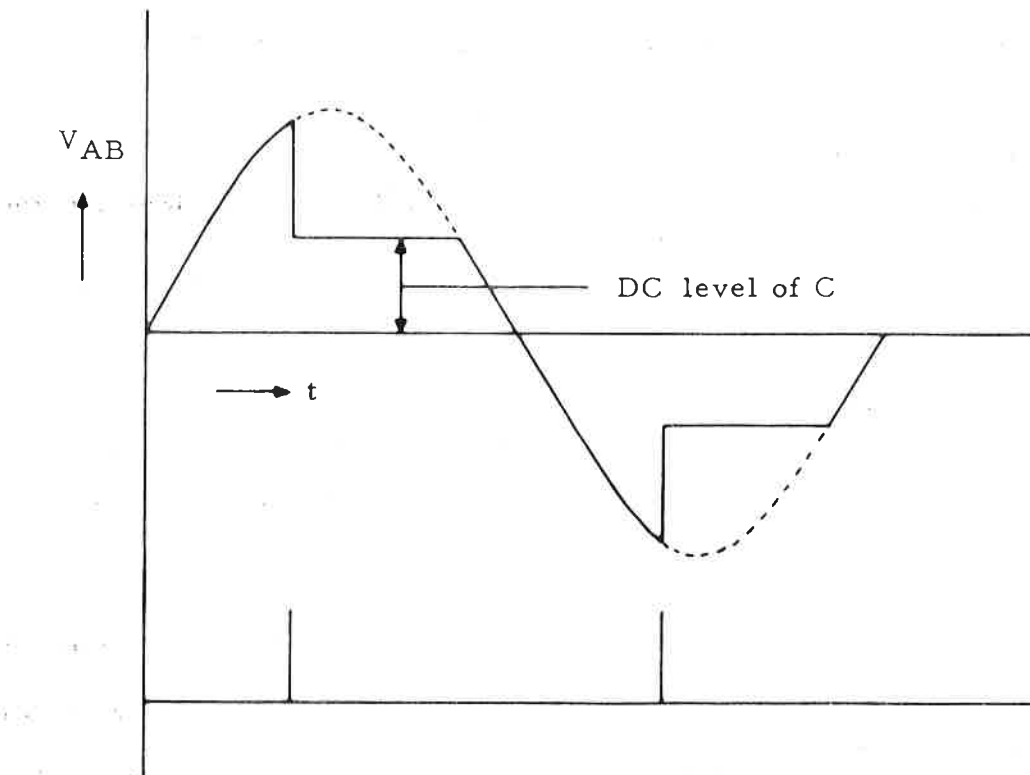


Fig. 3

The error voltage is converted into a time shift of the firing pulses of the controlled rectifiers, with regard to the zero crossings of the sine wave input voltage.

This causes a voltage change across the capacitor C in order to keep the voltage across T constant.

At a load current of 3 Amp. the voltage across T is about 3 Volts. So the dissipation of T is than 9 Watts.

Without pre-regulation the dissipation would be about 90 Watts at low output voltages.

Constant voltage regulation:

A comparison bridge is formed with the resistors R4, R5, the reference voltage and the output voltage.

After amplification the error voltage of the bridge drives transistor T. At the condition of balance the output voltage is practically proportional to R5.

As long as the constant voltage regulation is active, the constant current regulation is inoperative, because one of the transistors of the current error amplifier is blocked.

Constant current regulation:

At constant current regulation the voltage drop across R1, which is proportional to the output current, is compared with the part of the reference voltage across R3.

The error voltage is amplified and drives transistor T in such a way that the voltage across R1 is kept constant and this means that the output current is constant.

The constant current is adjustable with potentiometer R3.

As long as the constant current regulation is active, the constant voltage regulation is inoperative, because one of the transistors of the voltage error amplifier is blocked.

The position of the crossover point of constant voltage regulation and constant current regulation depends on the settings of the voltage and current controls.

In fig. 4 the crossover point is drawn.

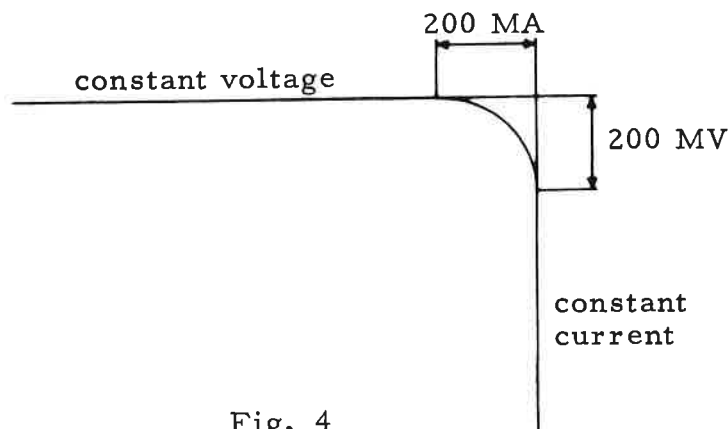


Fig. 4

R (Ohm)

1 =	1,2 k	5W 5% WW
2 =	330	
3 =	1 k	
4 =	33 k	
5 =	1,5 k	
6 =	1,2 k	MF
7 =	10	5W 5% WW
8 =	1,2 k	9W 5% WW
9 =	12 k	MF
10 =	4,7 k	MF
11 =	CR	
12 =	3,3 k	
13 =	47 k	
14 =	10	
15 =	100	
16 =	4,7 k	
17 =	1 k	
18 =	1 k	
19 =	1,2 k	9W 5% WW
20 =	330 k	
21 =	1 m	variable
22 =	1 k	
23 =	100	
24 =	0,68 k	25W 3% WW
25 =	1 k	variable
26 =	10 k	
27 =	8,2 k	
28 =	10 k	
29 =	100 k	
30 =	8,2 k	
31 =	3,9 k	
32 =	1,8 k	
33 =	330	
34 =	330	
35 =	1 k	variable
36 =	1 m	
37 =	15 k	
38 =	30 k	
39 =	6,8 k	$\frac{1}{2}$ W 2% WW
40 =	CR	
41 =	CR	
42 =	330	
43 =	CR	
44 =	82 k	MF
45 =	5 k	var. 10 t. potm.
46 =	5 k	var. 10 t. potm.
47 =	18 k	
48 =	6,8 k	MF
49 =	CR	
50 =	1 k	

C(microfarad)

1 =	100	100 V
2 =	0,07 + 2 x 0,0025	
	250 V- 50 Hz	F 1185/14 Roe
3 =	1	250 V
4 =	10	100 V
5 =	5900	70 V
6 =	25	70 V
7 =	25	15 V
8 =	250	15 V
9 =	250	15 V
10 =	0,01	250 V
11 =	0,01	250 V
12 =	10	100 V
13 =	10	100 V
14 =	0,47	250 V
15 =	0,1	630 V
16 =	500	70 V
17 =	100	70 V

D

1 =	TS 2	DI
2 =	VE 18	Varo
3,4,11 =	VTD 200/S	Varo
5 =	2N3668	RCA
6 =	2N3668	RCA
7 =	RE 38	Delta
8 =	ZP 6,8	Intermetall
9 =	TS 2	DI
10 =	ZP 6,2	Intermetall
12 =	ZP 6,8	Intermetall
13 =	OA 202	Philips

T

1 =	2N3053	RCA
2 =	BC 212	TI
3 =	BC 182	TI
4 =	BC 212	TI
5 =	BC 182	TI
6 =	2N3055	RCA
7 =	2N3055	RCA
8 =	2N3053	RCA
9 =	2N3053	RCA
10 =	BC 182	TI
11 =	BC 182	TI
12 =	BC 182	TI
13 =	BC 182	TI
14 =	2N3053	RCA

Ch = Choke S 030-3 Delta

Tr = Transformer T 030-3 D Delta

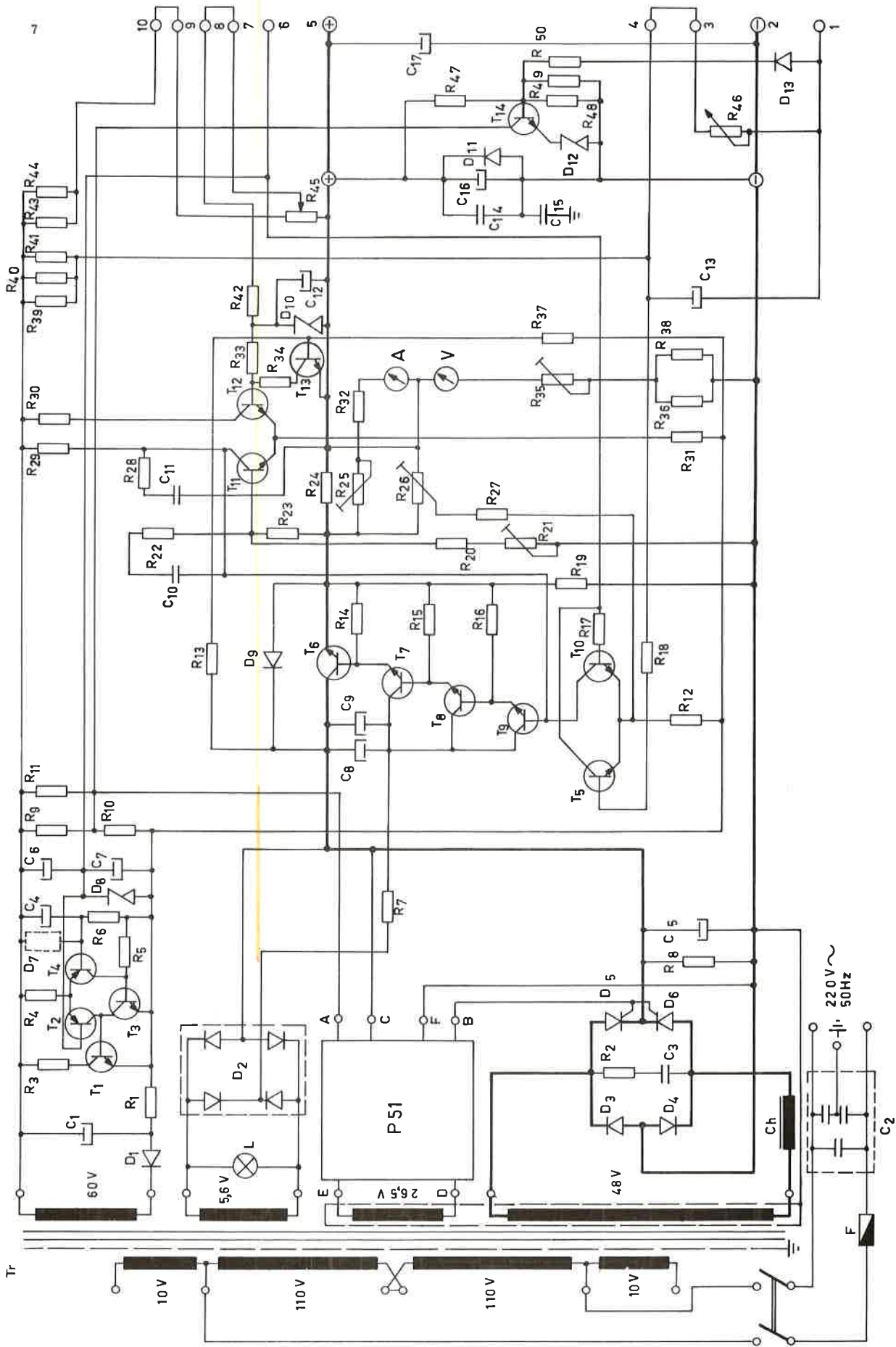
L = Liliput telefonlamp 6 V 0,04 A- Taunuslicht

F = Fuse 3,15 A -delay- $5\frac{1}{4}$ " x $\frac{1}{4}$ "- 250 V

CR = Calibration resistor

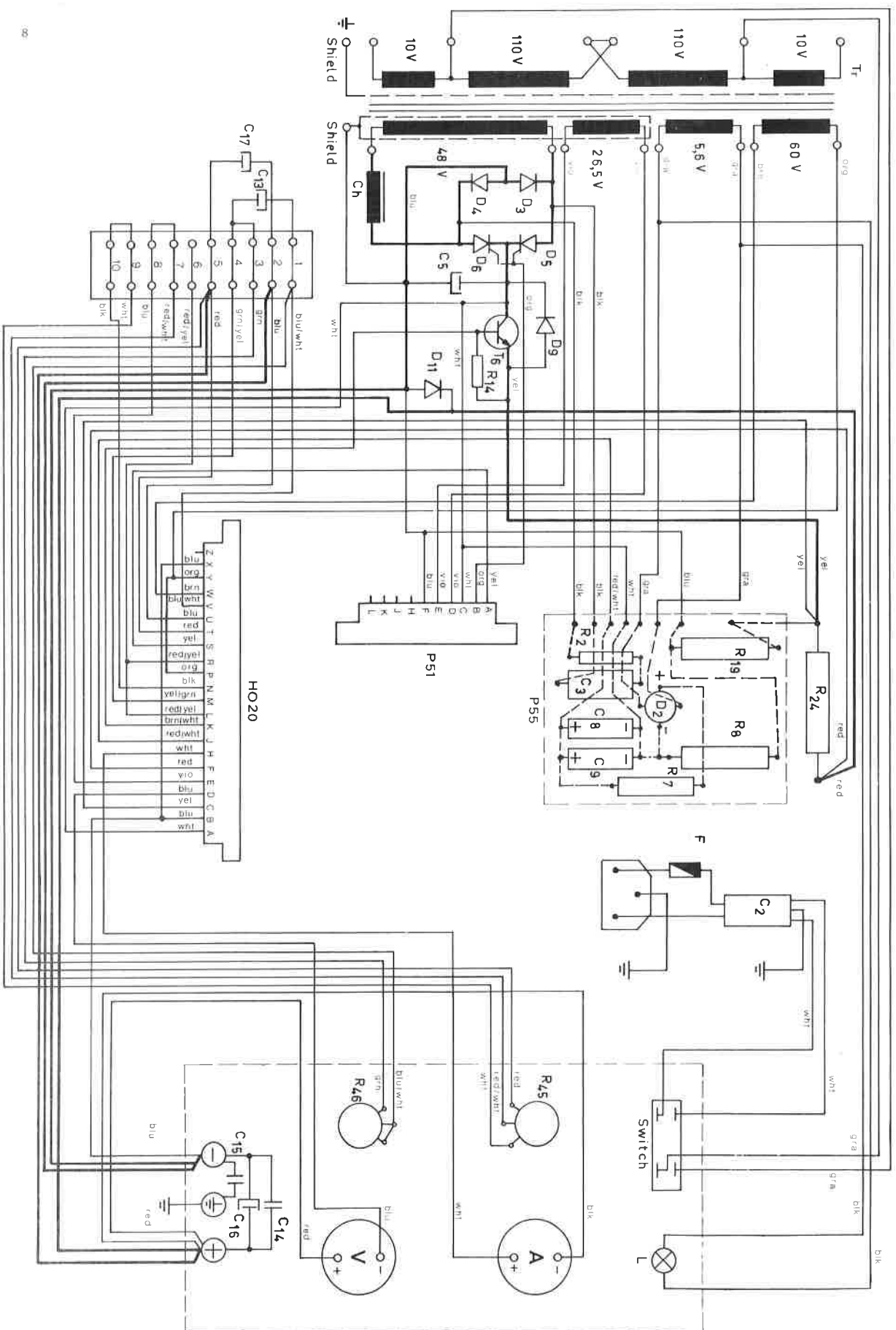
WW = Wire wound resistor

MF = Metalfilm resistor $\frac{1}{2}$ W 2%all other resistors carbon $\frac{1}{2}$ W 5%



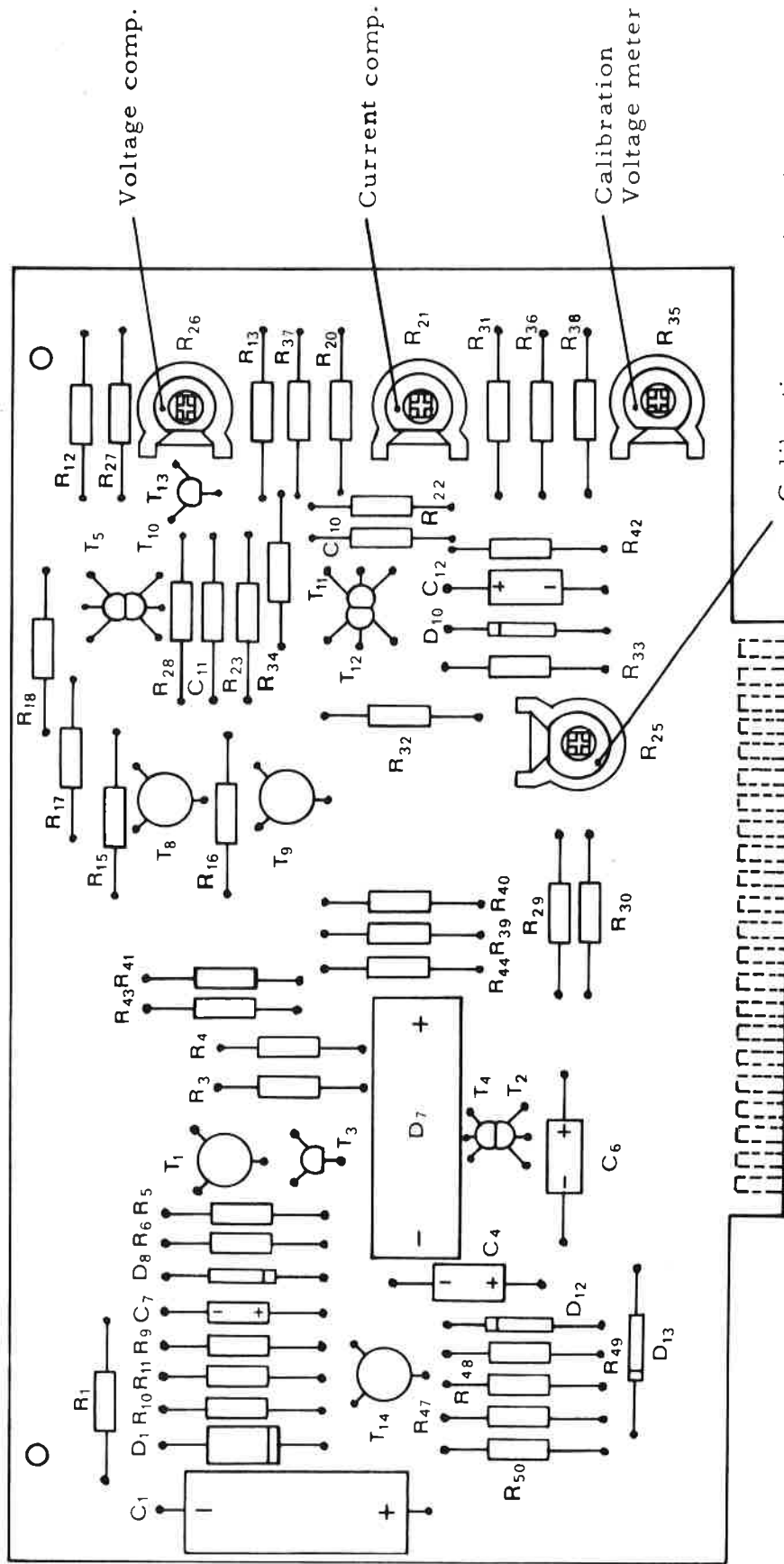
D030 - 3 , to start with serial nr 8173

jan.1969

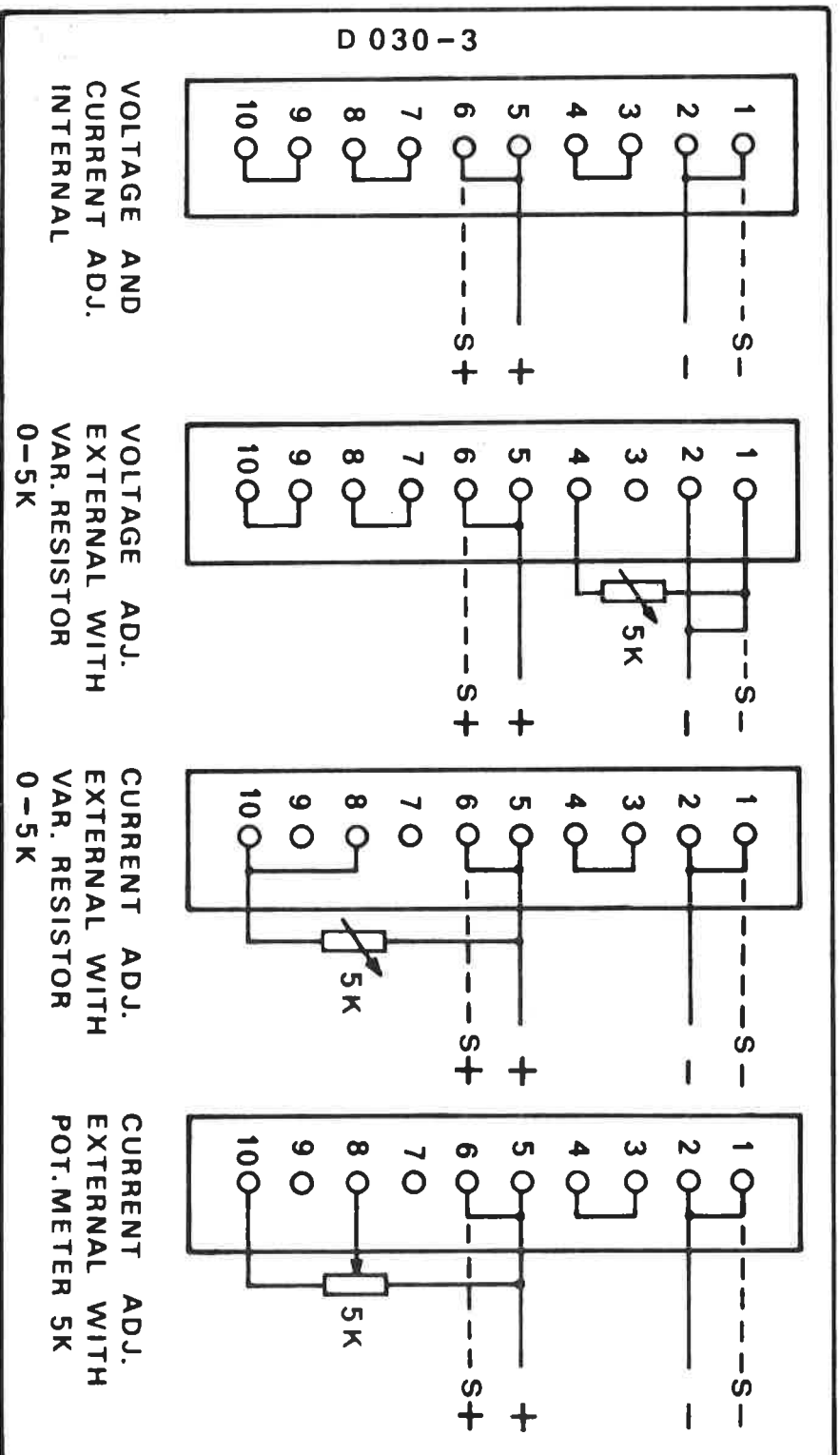


D030-3, to start with serial nr. 8173

Jan 1969



HO 20

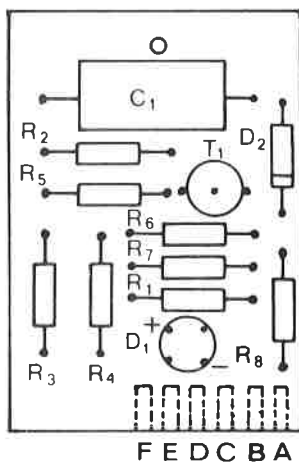
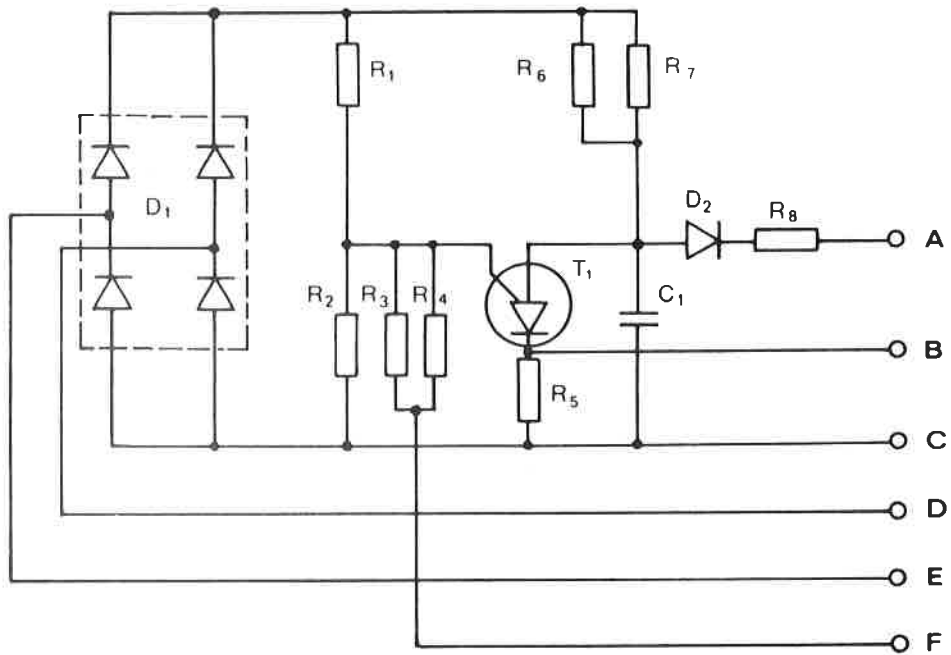


Barrier strip connections on rear of D 030-3

D 030-3, to start with serial nr 8173

DELTA ELEKTRONIKA

jan. '69



Pulse unit P 51

R (Ohm)

- 1 = 2,7 k $\frac{1}{2}$ W 2% MF
- 2 = 2,7 k $\frac{1}{2}$ W 2% MF
- 3 = CR
- 4 = 15 k $\frac{1}{2}$ W 2% MF
- 5 = 27 $\frac{1}{2}$ W 5%
- 6 = 15 k $\frac{1}{2}$ W
- 7 = CR
- 8 = 4,7 k $\frac{1}{2}$ W 2% MF

C (microfarad)

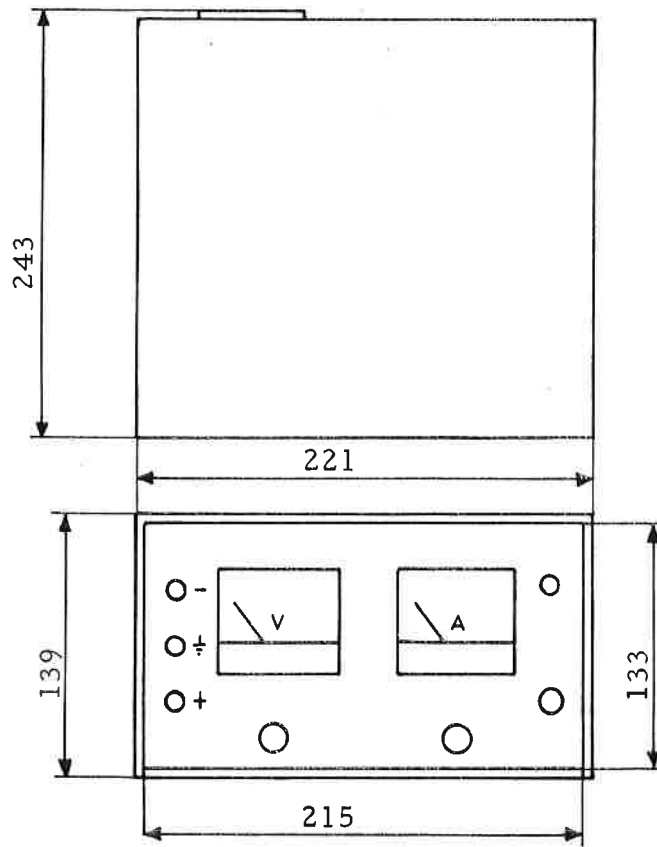
- 1 = 0,22 63 V 2%

D

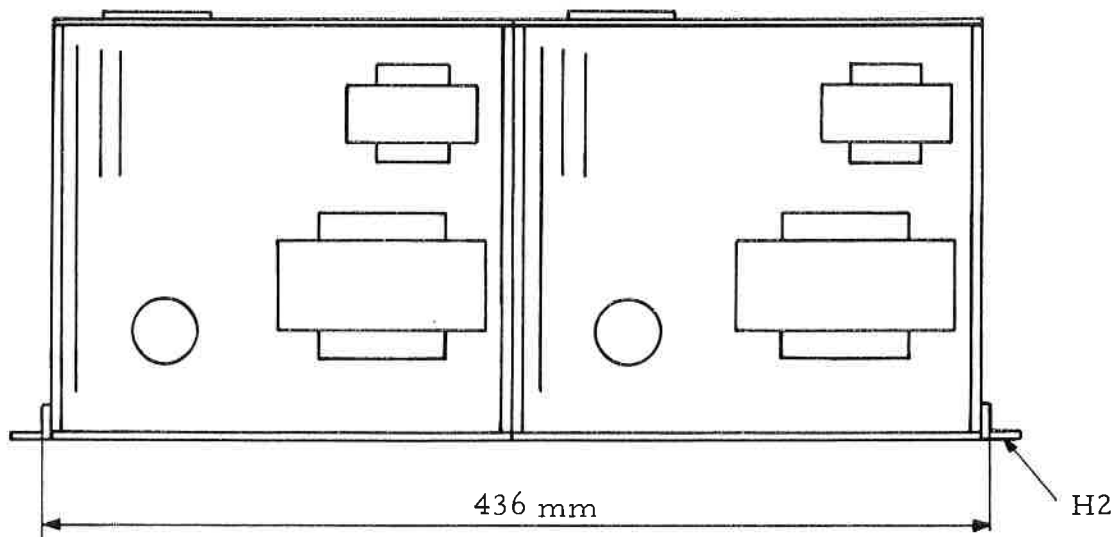
- 1 = W113 Varo
- 2 = OA202 Philips

T

- 1 = D13T1 G.E.



D 030-3 A



2 x D 030-3 B

